

The University of Manchester



Automated PCB Component Placement and Copper Trace Layout for Power Electronics

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About us

Design and Optimisation Automation for Power Electronics (DOAPE) Power Conversion Group, Department of Electrical and Electronic Engineering



The University of Manchester





Dr Gus Zhang (me) Lecturer

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Dr Judith Apsley Senior Lecturer





Miss Zhuoru Li PGR Student

About this presentation

It is about our <u>explorations</u> in

Design and optimisation automation of PCB layouts

for Power Electronics

and we share the *experiences*,

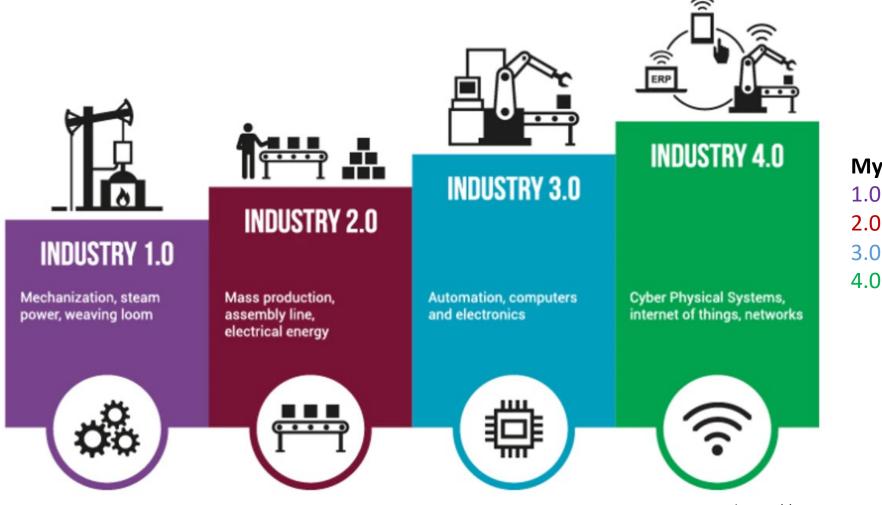
initiate the *discussions*,

tell what doesn't work and what may work, and look forward to future *collaborations*.

Outline

- Background, Aims and Objectives why we're doing this
- Methodologies what are the ideas
- Implementations what we have encountered
- Examples what we have done
- Conclusion next plans
- Q & A

Background – the trend



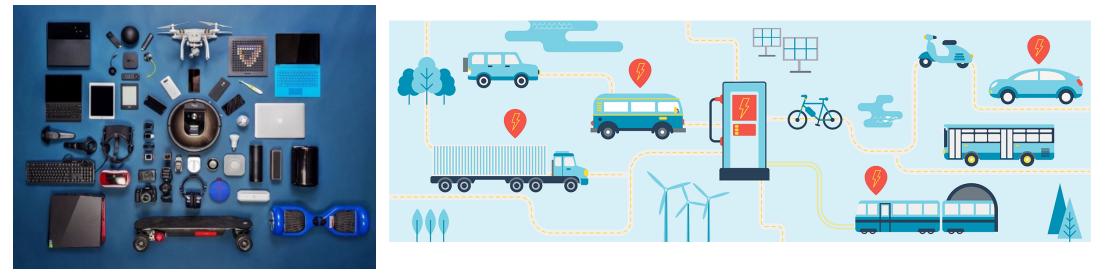
My interpretation:

1.0 - New power sources
2.0 - New production tools
3.0 - Automated production processes
4.0 and future -Automated design processes

Image: automationworld.com

The linkage to Power Electronics

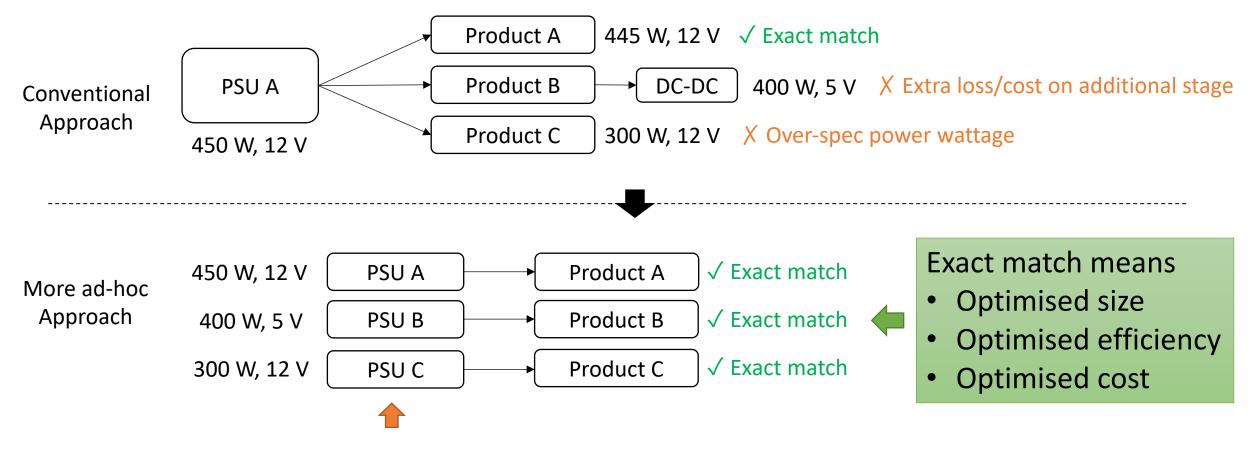
 Almost all electrical/electronic products have a part of power electronics – as long as there is energy conversion!



- Need to redesign them every time?
 - Need to optimise for efficiency, cost, volume and weight ...

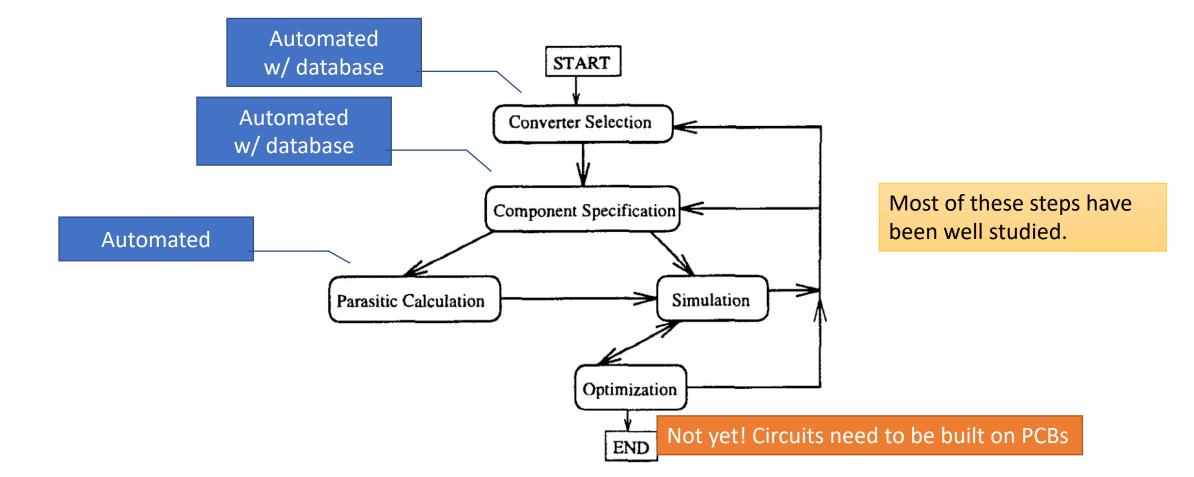
Source: consumertechtips.com, portlandgeneral.com

An emerging demand of more circuit designs



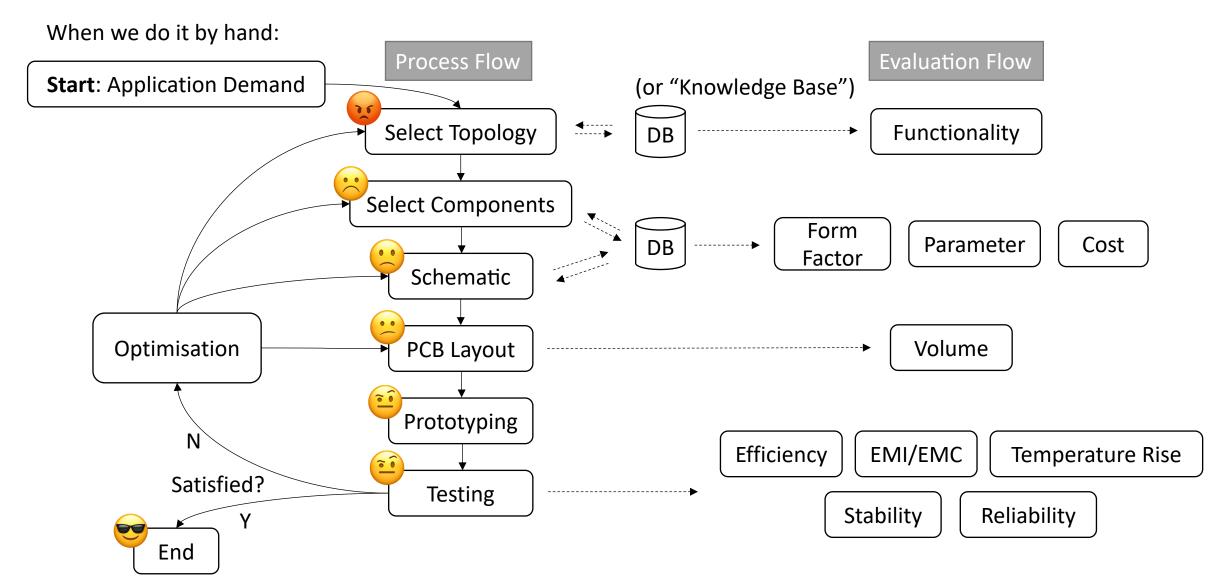
3x the work of design, development and testing!

Run-down of a generic PE design

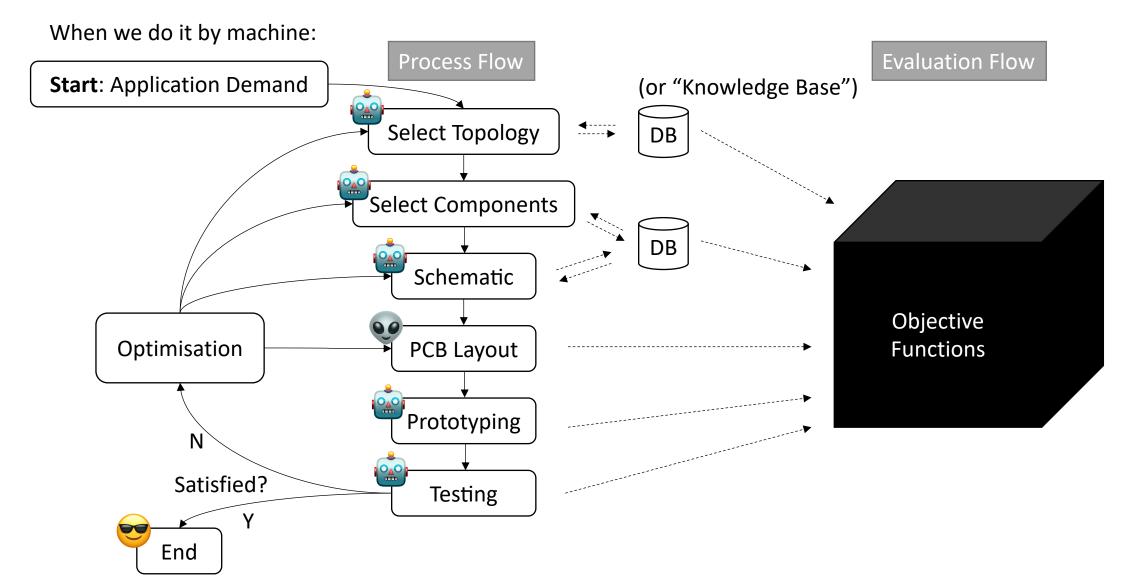


L. E. Amaya, P. T. Krein and F. N. Najm, "A design methodology for power electronics," 5th IEEE Workshop on Computers in Power Electronics, 1996, pp. 52-57, doi: 10.1109/CIPE.1996.612336.

A resketch, rough and messy...

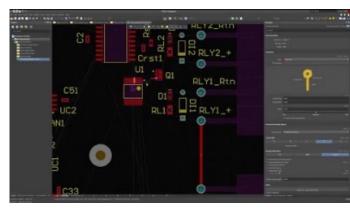


A resketch, rough and messy...

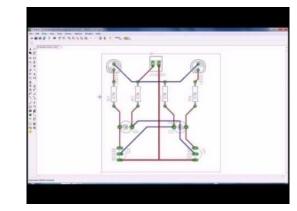


Existing auto-routing and auto-placement

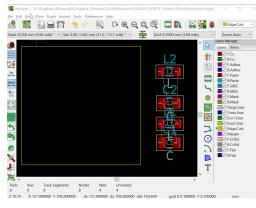
- Problems:
 - Simplified models assuming PCB tracks are ideal conductors
 - No or little considerations on component interconnections
 - Work for simple, low frequency circuits
 - Integrated calculators mainly for impedance matching and latencies → for high-speed mixed signal circuits



Altium Designer



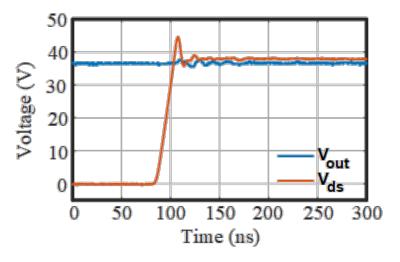
EAGLE



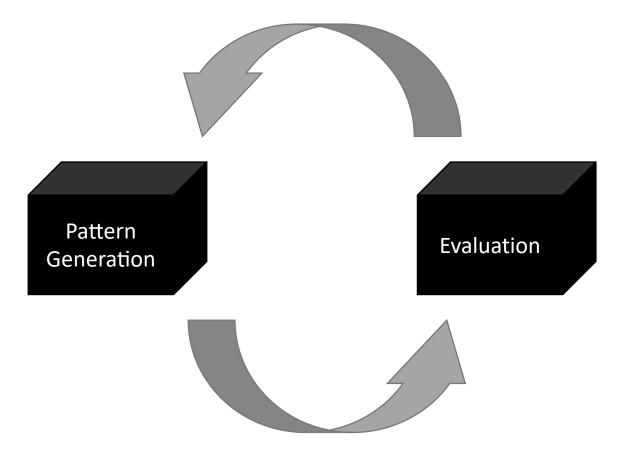


Unsuitable for Power Electronics!

- For power electronics we care about:
 - Energy efficiency!
 - Less I2R loss
 - Faster and clean dv/dt and di/dt edges
 - In other words, need traces to be as close as ideal
 - Box volume
 - Thermal dissipation (e.g. balanced distribution)



Method: Generative and Evaluative



The scope

- We target on PCB layout automation only
 - Automated placement
 - Automated trace layout, and copper pours
 - Automated via placement
 - CAM files generation
- In future developments we would like to integrate with other design automation tools
 - Topologies
 - Component selection
 - Control
 - ...

Assumptions

- Circuit topology is known
- Components, including values and packages are determined
- Circuit operation is known
 - Therefore, all node voltages and terminal currents are known, through SPICE simulation
- Targets in layman's words:
 - Least Ohmic loss on PCB
 - Least parasitic impact
 - Use intuition to define what is the best PCB layout

Implementation – an overview External Manual Process Input: netlist Input: footprint **Design Rules** DRC Human **Design Rules** LTSPICE Generator Input: waveforms Pad-to-pad Component Via Farm Output: Copper Placement Connection Pouring Filling Gerber Connection Weightings Hierarchical Grouping Parasitic Parameter Manual Evaluation Grouping

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Input: Netlist

- Widely used to represent circuit the machine readable schematic
- Used for both simulation and topology extraction

• Examples:

1	<pre>* C:\Users\Frank Tian\Desktop\DC-DC_design\boost_des</pre>
2	L1 IN N001 3300
3	R1 N002 N003 2
4	V1 IN 0 100
5	Rload OUT 0 3200
6	C1 OUT 0 0.270 Rser=370m
7	V2 N003 0 PULSE(-5 5 0 1e-9 1e-9 1.47e-6 2e-6)
8	XU2 N001 OUT C3D1P7060Q
9	XU1 N002 N001 0 0 GaN_LTspice_GS66504B_L1V4P1
10	.tran 2m startup
11	.lib C:\Users\Frank Tian\Desktop\DC-DC_design\LTspic
12	.lib GaN_LTspice_GS66504B_L1V4P1.lib
13	.backanno

signal name="VSN SENSE" <contactref element="R3" pad="2"/> <contactref element="R2" pad="1"/> <contactref element="C5" pad="1"/> <contactref element="C4" pad="2"/> <wire x1="20.75" y1="12.9" x2="21.65" y2="12.9" width="0.4" layer="1"/> <wire x1="20.75" y1="11.9" x2="21.65" y2="11.9" width="0.4" layer="1"/> <wire x1="20.75" y1="11.9" x2="20.9" y2="11.9" width="0.4" layer="1"/> <wire x1="20.9" y1="11.9" x2="21.2" y2="12.2" width="0.4" layer="1"/> <wire x1="21.2" y1="12.2" x2="21.2" y2="12.7" width="0.4" layer="1"/> <wire x1="21.2" y1="12.7" x2="21.45" y2="12.7" width="0.4" layer="1"/> <wire x1="21.45" y1="12.7" x2="21.65" y2="12.9" width="0.4" layer="1"/> <wire x1="20.75" y1="11.9" x2="21.7" y2="11.9" width="0.4" layer="1"/> <wire x1="21.7" y1="11.9" x2="21.8" y2="11.8" width="0.4" layer="1"/> <wire x1="21.8" y1="11.8" x2="21.8" y2="10.9" width="0.4" layer="1"/> <via x="21.8" y="10.9" extent="1-16" drill="0.3" diameter="0.4"/> <wire x1="21.8" y1="10.9" x2="21.8" y2="7.9" width="0.4" layer="15"/> <wire x1="21.8" y1="7.9" x2="19.7" y2="5.8" width="0.4" layer="15"/> <via x="19.7" y="5.8" extent="1-16" drill="0.3" diameter="0.4"/> <contactref element="JP1" pad="10"/> <wire x1="19.7" y1="5.8" x2="18.7" y2="4.8" width="0.4" layer="1"/> <wire x1="18.7" y1="4.8" x2="18.7" y2="3.88" width="0.4" layer="1"/>

.brd file from EAGLE

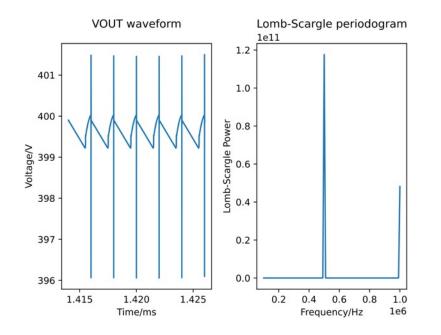
.net file for SPICE

Simulation result extraction

LTSPICE outputs .raw waveform files includes all

- Voltages on nodes
- Currents on component terminals

Unfixed step-size: Lomb-Scargle periodogram to find out the period



Title: * C:\Users\Frank Tian\Desktop\DC-DC_design\boost_design\boost_converter_50W.asc
Date: Tue Sep 07 11:03:52 2021
Plotname: Transient Analysis
Flags: real forward
No. Variables: 18
No. Points: 1156432
Offset: 0.00000000000000000000000000000000000
Command: Linear Technology Corporation LTspice XVII
Backannotation: u2 a k
Backannotation: ul gatein drainin sourcein source_s
Variables:
0 time time
1 V(in) voltage
2 V(n001) voltage
3 V(n002) voltage
4 V(n003) voltage
5 V(out) voltage
6 I(Cl) device_current
7 I(L1) device_current
8 I(Rload) device_current
9 I(R1) device_current
10 I(V2) device_current
11 I(V1) device_current
12 Ix(u2:A) subckt_current
13 Ix(u2:K) subckt_current
14 Ix(u1:GATEIN) subckt_current
15 Ix(ul:DRAININ) subckt_current
16 Ix(ul:SOURCEIN) subckt_current
17 Ix(u1:SOURCE_S) subckt_current
Binary:
¢0¢0¢0□0側¢間上□阖貫なカ□⊍構兀構兀恭났□⊍□⊳ 恭袒女カ□风=ゝ」□・摘好\$陣挥戟偿鬻뢳敳□嗄犖埭湊才說□□岁□왕湊낪尋
蓝፲頊髦質←習∈製ψ劑煵㽏恀獲藕獲ሧ應Δ搓□尲藕뻸腱獲腱뺈Δ랉寏燥ΰ計镥뤿Ӥ最↔珛꽡質射習コフ製ψ凹□甞潸獲毬獲□懦廾装

We fetch:

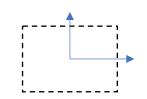
- Maximum absolute value and RMS value of all node pairs voltages, and their derivatives
- Maximum absolute value and RMS value of all terminal currents, and their derivatives

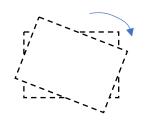
Simulation result extraction

- For voltages between any pair of nodes (nets)
- $v_{max} \rightarrow$ related to clearance distances
- $dv/dt \rightarrow$ related to stray capacitance, need to be minimised
- *n* nodes $\rightarrow n(n-1)/2$ voltage differences
- For current between any pair of terminals (pads) "connections"
- i_{max} , i_{rms} \rightarrow related to I2R losses
- $di/dt \rightarrow$ related to impacts from inductances

Placement solution space

- For each component: origin offset (x, y) and angle (θ) , all $\in \Re$
 - Regardless of the shape of the component





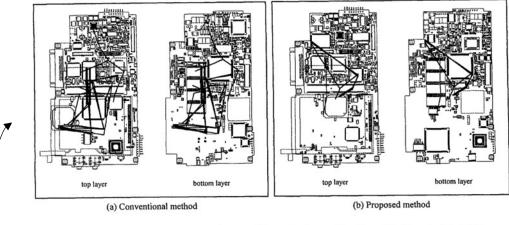


Figure 7. Comparison between proposed automatic layout and conventional layout for the main board of a multimedia equipment.

Ref & most CAD: only limited to orthogonal placement

In practice:

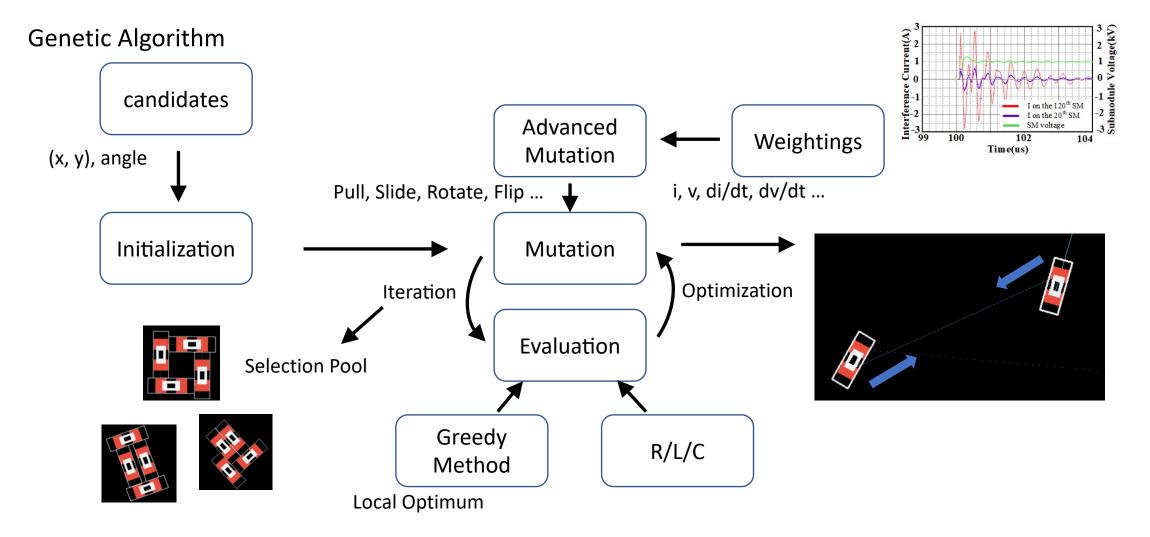




• Still limited degree-of-freedom, 3x (component count-1)

Y. Fukumoto, S. Miura, H. Ikeda, T. Nakayama, S. Tanimoto and H. Uemura, "A method of automatic placement that reduces electromagnetic radiation noise from digital printed circuit boards," IEEE International Symposium on Electromagnetic Compatibility. Symposium Record (Cat. No.00CH37016), Washington, DC, 2000, pp. 363-368 vol.1, doi: 10.1109/ISEMC.2000.875594.

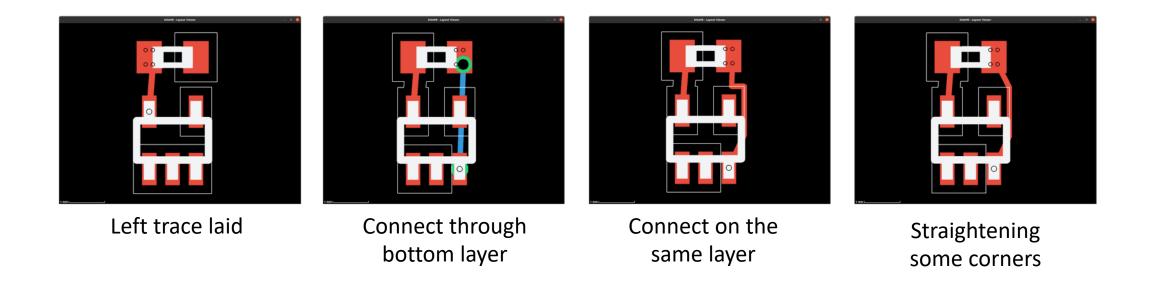
Implementation – Component placement



Connecting pads with traces

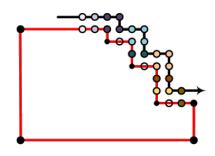
Bug algorithm – walk around obstacles

A* search algorithm – existing and predicted cost weightings via key points

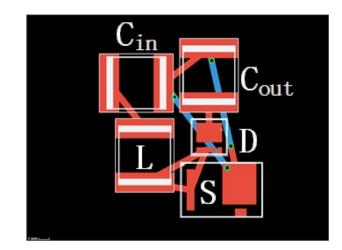


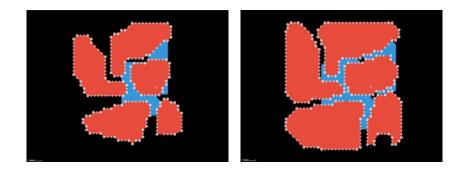
Copper pouring

Rasterised approach

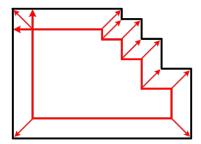


Acceptable resolution and execution efficiency, easy to control speed of expansion.

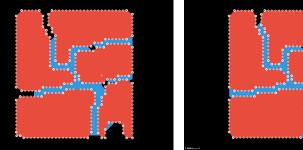


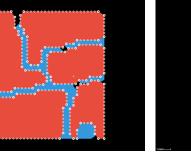


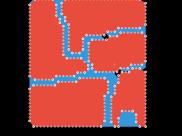
Vector approach



Very complicated computation near narrow and sharp corners.



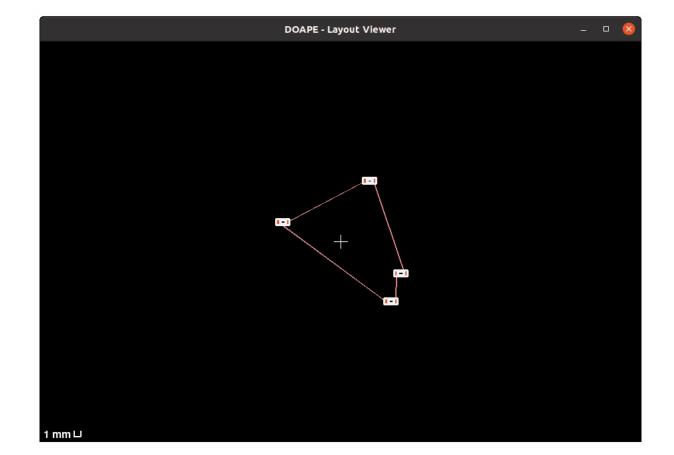


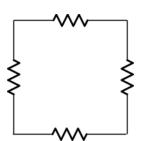


After pouring, the rasterised copper areas will be converted back to the vector formats.

Examples – Placement

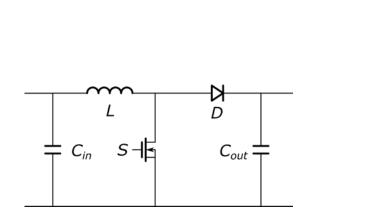
A simple case with four chip resistors head-to-head

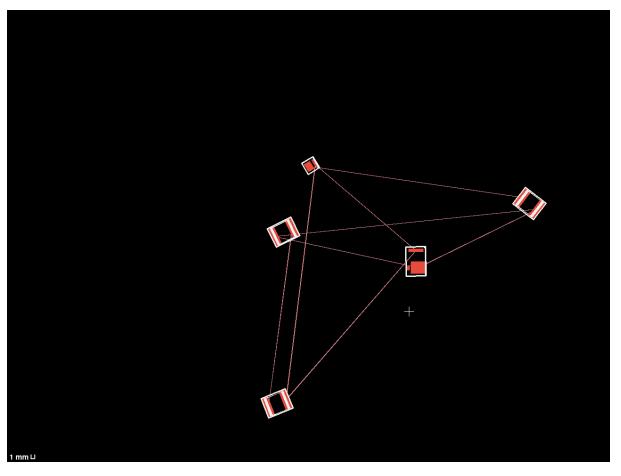




Examples – Placement and Copper Pouring

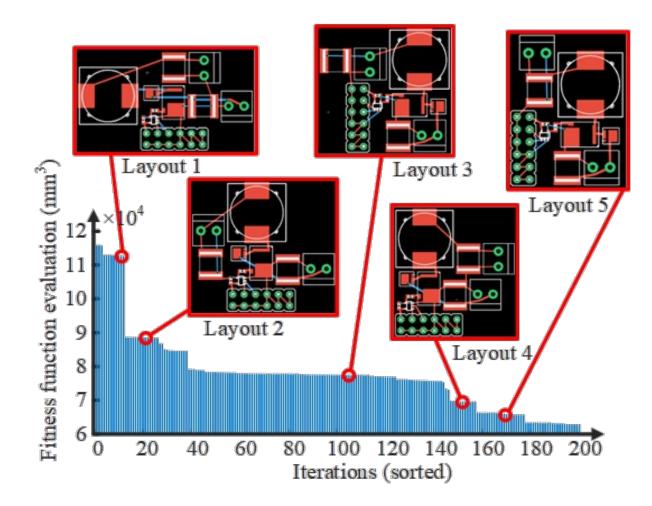
A simple case with five major components in a boost converter.

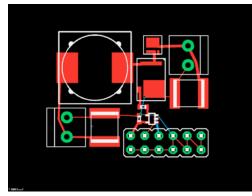




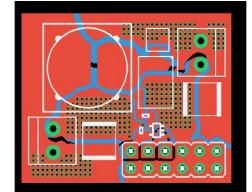
Example: A Simple Boost Converter

A more practical example: a boost converter with connectors and gate driver

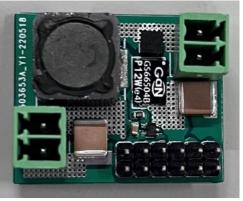




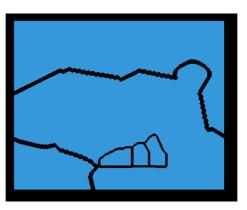
Placement



Layout (top layer)

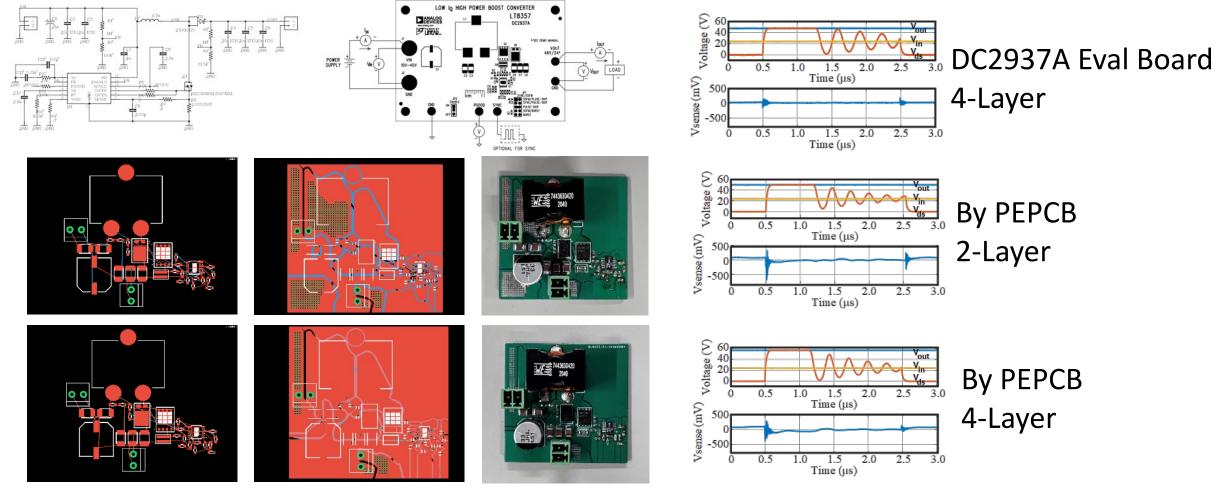


Product



Layout (bottom layer)

Example: LT2937A Boost Converter



2.5x smaller than eval board, 30 minutes on i7-10510U+16G RAM

Conclusion

- We have explored automated PCB layout design including the steps of
 - Extraction of circuit topology and voltage/current information
 - Synthesising grouping hierarchy and weightings
 - Automatic placement subject to given rules and objectives
 - Automatic trace connections and copper pouring
- Big picture is still genetic algorithm, but lots of the works are done by hard-coding for better efficiency
- We're still improving all steps through testing with more complicated circuits. Many techniques are constantly being evolved.

Additional notes

- We have used Python for generation most of the examples in this presentation. We are migrating the algorithms to C++ for better performance and more rigorous pre-compile checks.
- We have used floating number for vector coordinates. We will move to integers to avoid boundary ambiguity problems.
- Limited by the length of the presentation, we cannot include all details. Please feel free to contact us!

Thank you!

For any queries please contact cheng.zhang@manchester.ac.uk